

Quasi-Monte Carlo Time-Frequency Analysis

Barbara Pascal

May 12, 2021

from R. Levie, H. Avron, and G. Kutyniok [arxiv:2011.02025](https://arxiv.org/abs/2011.02025)

But first ... music!

- Born in Leeds, in 1946
- Education in music
 - Oxford (BA, '68)
 - Nottingham (MA, '69)

But first ... music!

- Born in Leeds, in 1946
- Education in music
 - Oxford (BA, '68)
 - Nottingham (MA, '69)
- † death of his father (factory worker) '69



Abandoned *conventional* composing

- ▶ collected sounds of machinery, foundries and power stations
- ▶ thesis on sonic art at York (PhD, '73)
- ▶ interaction of human voice and electronic systems

But first ... music!

- Born in Leeds, in 1946
- Education in music
 - Oxford (BA, '68)
 - Nottingham (MA, '69)
- † death of his father (factory worker) '69



Abandoned *conventional* composing

- ▶ collected sounds of machinery, foundries and power stations
- ▶ thesis on sonic art at York (PhD, '73)
- ▶ interaction of human voice and electronic systems

<http://www.trevorwishart.co.uk/>

Vox 5, '86-'87

Transformation and interpolation with natural sounds of human voice

Institut de Recherche et Coordination Acoustique/Musique-Centre Pompidou

Elementary transforms used in time-frequency analysis

Action in the time domain $s \in L^2(\mathbb{R})$

- translation by x : $[\mathcal{T}(x)s](t) = s(t - x)$
- modulation by ω : $[\mathcal{M}(\omega)s](t) = s(t)e^{2i\pi\omega t}$
- dilation by τ : $[\mathcal{D}(\tau)s](t) = \tau^{-1/2}s(\tau^{-1}t)$

Elementary transforms used in time-frequency analysis

Action in the time domain $s \in L^2(\mathbb{R})$

- translation by x : $[\mathcal{T}(x)s](t) = s(t - x)$
- modulation by ω : $[\mathcal{M}(\omega)s](t) = s(t)e^{2i\pi\omega t}$
- dilation by τ : $[\mathcal{D}(\tau)s](t) = \tau^{-1/2}s(\tau^{-1}t)$

Lemma [Unitarity] For any $x \in \mathbb{R}, \omega \in \mathbb{R}, \tau \in \mathbb{R} \setminus \{0\}$,

$\mathcal{T}(x)$, $\mathcal{M}(\omega)$ and $\mathcal{D}(\tau)$ are *unitary* operators of $L^2(\mathbb{R})$

$\mathcal{F} : L^2(\mathbb{R}) \rightarrow L^2(\mathbb{R})$ the Fourier transform

Action in the frequency domain

- $\mathcal{F}\mathcal{T}(x)\mathcal{F}^* = \mathcal{M}(-x)$
- $\mathcal{F}\mathcal{M}(\omega)\mathcal{F}^* = \mathcal{T}(\omega)$
- $\mathcal{F}\mathcal{D}(\tau)\mathcal{F}^* = \mathcal{D}(\tau^{-1})$

Short-Time Fourier and Continuous Wavelet Transforms

STFT

$$f_{a,b} = \mathcal{T}(a)\mathcal{M}(b)f$$

$f \in L^2(\mathbb{R})$ window localized in time and frequency about 0

Fixed length of support
Varying nb. of oscillations

The larger the frequency, the more oscillations.

CWT

$$f_{a,b} = \mathcal{T}(a)\mathcal{D}(b^{-1})f$$

f mother wavelet localized in time about 0, in frequency about 1

$$\int \omega |\widehat{f}(\omega)|^2 d\omega < \infty \text{ (admissibility)}$$

Varying length of support
Fixed nb. of oscillations

The larger the frequency, the shorter the time support.

Localizing Time-Frequency Transform (LTFT)

Heuristic

- ▶ Short-Time Fourier atoms at low and high frequencies (STFT)
- ▶ Continuous Wavelet atoms for middle frequencies (CWT)
- ▶ *Third* parameter: number of oscillations in the *mother* wavelet.

Localizing Time-Frequency Transform (LTFT)

Heuristic

- ▶ Short-Time Fourier atoms at low and high frequencies (STFT)
- ▶ Continuous Wavelet atoms for middle frequencies (CWT)
- ▶ *Third* parameter: number of oscillations in the *mother* wavelet.

Definition [LTFT continuous frame] $f_{a,b,c} = \pi(a, b, c)f$ with

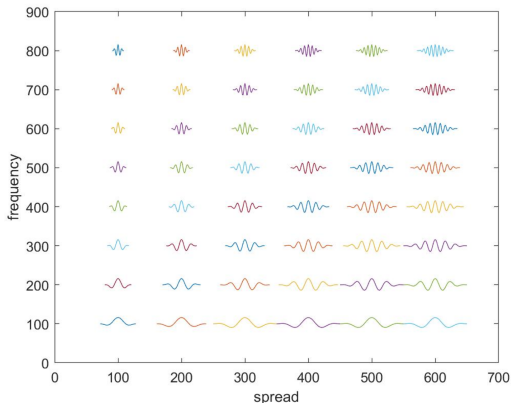
$$\pi(a, b, c) = \begin{cases} \mathcal{T}(a)\mathcal{M}\left(\frac{\xi}{\gamma}cb_0 + b\right)\mathcal{D}\left(\frac{\gamma}{b_0}\right) & b < b_0 \\ \mathcal{T}(a)\mathcal{M}\left(\left(\frac{\xi}{\gamma}c + 1\right)b\right)\mathcal{D}\left(\frac{\gamma}{b}\right) & b_0 \leq b \leq b_1 \\ \mathcal{T}(a)\mathcal{M}\left(\frac{\xi}{\gamma}cb_1 + b\right)\mathcal{D}\left(\frac{\gamma}{b_1}\right) & b > b_1 \end{cases}$$

- “father” atom: $f \in L^2(\mathbb{R})$ support. in $[-1/2, 1/2]$ localized at 0 in T & F
- phase space: $G = \mathbb{R}^2 \times [0, 1]$ (time, frequency, nb. of oscillations)
- STFT-CWT transition frequencies: $0 < b_0 < b_1$
- minimal nb. of oscillations: γ
- oscillation range: ξ

Localizing Time-Frequency Transform (LTFT)

Heuristic

- ▶ Short-Time Fourier atoms at low and high frequencies (STFT)
- ▶ Continuous Wavelet atoms for middle frequencies (CWT)
- ▶ *Third* parameter: number of oscillations in the *mother* wavelet.



Theory of *frames*

- \mathcal{S} : Hilbert space of signals
e.g.: finite energy signals $L^2(\mathbb{R})$
- G : locally compact Borel space with measure μ
e.g.: time-frequency plane \mathbb{R}^2 with Lebesgue measure λ

Definition [Continuous frame]

Frame $f : \begin{cases} G & \rightarrow \mathcal{S} \\ g & \mapsto f_g \end{cases}$ weakly-measurable mapping

$V_f[s] : \begin{cases} G & \rightarrow \mathbb{C} \\ g & \mapsto \langle s, f_g \rangle_{\mathcal{S}} \end{cases}$ the *coefficient function* of signal $s \in \mathcal{S}$.

Continuous $\forall s \in \mathcal{S}, V_f[s] \in L^2(G)$ and $\exists A, B, 0 < A \leq B < \infty$, s.t.

$$A\|s\|_{\mathcal{S}}^2 \leq \|V_f[s]\|_{L^2(G)}^2 \leq B\|s\|_{\mathcal{S}}^2.$$

V_f *analysis* operator, V_f^* *synthesis* operator

Signal processing in phase space

General procedure

- (i) compute $V_f[s](g)$ for every point of phase space $g \in G$
- (ii) apply a nonlinearity $\kappa(V_f[s](g), g)$
- (iii) map each $g \in G$ onto $\rho(g)$
- (iv) synthesize the output signal

$$s_{\text{out}} = \int_G \kappa(V_f[s](g), g) f_{\rho(g)} d\mu(g)$$

Examples

- multipliers: $\kappa(c, g) = cr(g)$, $\rho(g) = g$
- signal denoising: $\kappa(c, g) = \kappa(c)$, $\rho(g) = g$
- time stretching vocoder: V_f Short-Time Fourier Transform

“slow down a signal without dilating its frequency content”

$g = (t, \omega)$, $\rho(g) = (Dt, \omega)$ for some $D \in \mathbb{N}$, $\kappa(c, g) = \kappa(c)$ and

$$\forall a, \theta \in \mathbb{R}_+, \quad \kappa(e^{i\theta} a) = e^{iD\theta} a$$

Notations M : resolution of the signal space,
 N : number of samples in phase space
 d : dimension of phase space.

- $\mathcal{O}\left(\frac{M}{N}(\log_2 N)^{d-1}\right)$ error rate of proposed quasi-Monte Carlo LTFT
 compared to $\mathcal{O}\left(\sqrt{\frac{M}{N}}\right)$ for standard Monte Carlo
- implementation of a *time dilation phase vocoder* based on LTFT
- general theory for quasi-Monte Carlo sampling of phase spaces
 STFT, CWT, LTFT, Shearlet or Curvelet transforms

Let $f : I^d = [0, 1]^d \rightarrow \mathbb{C}$, QMC is a cubature method to compute

$$\int_{I^d} f(t) \, dt \simeq \frac{1}{N} \sum_{n=1}^N f(x_n)$$

from samples points $\mathcal{P}_n = \{x_1, \dots, x_N\}$

Let $f : I^d = [0, 1]^d \rightarrow \mathbb{C}$, QMC is a cubature method to compute

$$\int_{I^d} f(t) dt \simeq \frac{1}{N} \sum_{n=1}^N f(x_n)$$

from samples points $\mathcal{P}_N = \{x_1, \dots, x_N\}$

Star discrepancy $D^*(\mathcal{P}_N) = \sup_{B \in \text{Rec}^*} \left| \frac{\#(B \cap \mathcal{P}_N)}{N} - \lambda(B) \right|$

how well the volume is covered

Quasi-Monte Carlo

Multi-index $\alpha = (\alpha_1, \dots, \alpha_d)$, $\alpha_j \in \{0, 1\}$ and $g : I^d \rightarrow \mathbb{C}$

$$\int_{I^d|_\alpha} g(a_1, \dots, a_d) \, da^\alpha$$

- ▶ integration w.r.t. a_j if $\alpha_j = 1$
- ▶ replace a_j by 1 if $\alpha_j = 0$

Ex.
$$\int_{I^3|_{(1,0,1)}} g(a_1, a_2 a_3) \, da^{(1,0,1)} = \int_{[0,1]^2} g(a_1, 1, a_3) \, da_1 da_3$$

Hardy-Krause variation
$$V(f) = \sum_{\alpha} \int_{I^d|_\alpha} |\partial_\alpha f(a_1, \dots, a_d)| \, da^\alpha$$

Theorem [Koksma-Hlawka inequality]

$$\left| \int_{I^d} f(x) \, dx - \frac{1}{N} \sum_{n=1}^N f(x_n) \right| \leq V(f) D_N^*(\mathcal{P}_N)$$

Quasi-Monte Carlo

Multi-index $\alpha = (\alpha_1, \dots, \alpha_d)$, $\alpha_j \in \{0, 1\}$ and $g : I^d \rightarrow \mathbb{C}$

$$\int_{I^d|_{\alpha}} g(a_1, \dots, a_d) da^{\alpha}$$

- ▶ integration w.r.t. a_j if $\alpha_j = 1$
- ▶ replace a_j by 1 if $\alpha_j = 0$

Ex.
$$\int_{I^3|_{(1,0,1)}} g(a_1, a_2 a_3) da^{(1,0,1)} = \int_{[0,1]^2} g(a_1, 1, a_3) da_1 da_3$$

Hardy-Krause variation
$$V(f) = \sum_{\alpha} \int_{I^d|_{\alpha}} |\partial_{\alpha} f(a_1, \dots, a_d)| da^{\alpha}$$

Theorem [Koksma-Hlawka inequality]

$$\left| \int_{I^d} f(x) dx - \frac{1}{N} \sum_{n=1}^N f(x_n) \right| \leq V(f) D_N^*(\mathcal{P}_N)$$

Linear volume discretizable frames

Let f a frame on the signal space \mathcal{S} with phase space G

Discretization of the frame:

- ▶ discrete signals $\{V_M \subset \mathcal{S}, \dim(V_M) = M \in \mathbb{N}\}$
- ▶ compact phase space $G_M \subset G$ rectangle of volume $\mathcal{O}(M)$

Sampling discretization of phase space: $\mathcal{P}_N = (g_1, \dots, g_N)$

ideally low discrepancy, i.e. $D_N^*(\mathcal{P}_N)$ “small”

Phase space signal processing via Quasi-Monte Carlo discretization

$$s_{\text{out}}^N = \frac{\mu(G_M)}{N} \sum_{n=1}^N \kappa(V_f[s](g_n), g_n) f_{\rho(g_n)}$$

$$s_{\text{out}} = \int_G \kappa(V_f[s](g), g) f_{\rho(g)} d\mu(g)$$

Sampling the phase space

Monte Carlo sampling

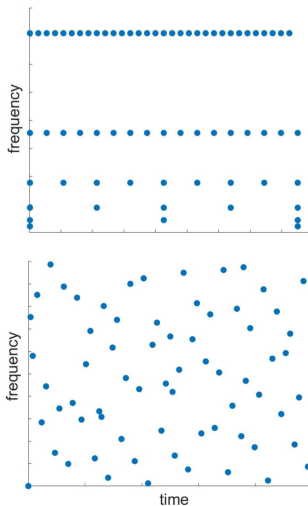
- well spread samples
- required nb. of samples ind. of d

no curse of dimensionality

- suitable for TF feature extraction

Quasi-Monte Carlo

- midway between grid and MC
- *optimally* spread



How to design a sampling strategy?

Reminder: purpose is to describe signals with a family of atoms

$$\{f_{g_n}, n = 1, \dots, N\}$$

How to design a sampling strategy?

Reminder: purpose is to describe signals with a family of atoms

$$\{f_{g_n}, n = 1, \dots, N\}$$

Requirements:

(i) well-spread samples

analysis extracts properly the local frequencies

(ii) allow reconstruction up to some small error

small information loss

(iii) sampling set small enough

computational efficiency

Standard discrete time-frequency transforms:

(ii) and (iii) satisfied ✓

(i) not satisfied ✗

Computational complexity

N : number of time-frequency samples

[*Total* number of pixels of *all* LTFT atoms] $\mathcal{C} = \mathcal{O}(\gamma N + \gamma)$

Error bound

M : resolution of the input discrete signal

Theorem [Approximation error of QMC LTFT]

$$\mathcal{O}\left(\frac{M \log^2(N)}{N}\right)$$

Approximation error of general QMC

M : resolution of the input discrete signal

d : dimension of phase space

N : number of time-frequency samples

QMC sample set $\mathcal{P}_N = \{(\mathbf{a}^n, \mathbf{b}^n), n = 1, \dots, N\}$ of compact phase space

with asymptotically *low* discrepancy $D_N^*(\mathcal{P}_N) \leq C \frac{(\log N)^{d-1} M}{N}$

Theorem [Reconstruction error] For $S : G_{M'} \rightarrow \mathbb{C}$ coefficients of a generalized TF transform, let $V_f^{M,N}(S) = \frac{M'}{N} \sum_{n=1}^N S(\mathbf{a}^n, \mathbf{b}^n) \underbrace{\mathcal{T}(\mathbf{a})f_b}_{\text{atoms}}$.

Then, for any discrete signal s_M of resolution M

$$\left\| V_f^{M*} V_f^M(s_M) - V_f^{M,N*} V_f^M(s_M) \right\|_{\infty} \leq \|s_M\|_{\infty} C \frac{(\log N)^{d-1} M}{N} D$$

► uniform pointwise bound on the QMC *synthesis* error.

Play your own music!

<https://github.com/RonLevie/LTFT-Phase-Vocoder>